

A PRACTICAL GUIDE TO UNDERSTANDING BEARING DAMAGE RELATED TO PWM DRIVES

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Abstract – The performance and reliability of AC Adjustable Speed Drives (ASD's) is continually improving. One of the key reasons for improvement has been the advent, development and use of Pulse Width Modulated (PWM) drives utilizing faster switching devices, primarily Insulated Gate Bipolar Transistors (IGBT's). As with many other developments, improvements in some areas may cause problems in others. An increased bearing failure rate in motors is one of the negative effects of these types of drives.

To mitigate bearing current damage in motors, as well as loads and other auxiliary equipment attached to the motor shaft, it is important to understand how these currents are generated.

In addition to theoretical explanations, actual field cases and solutions of how to minimize bearing damage will be reviewed.

1. INTRODUCTION

The phenomenon of motor shaft voltages producing circulating shaft currents has been recognized since the 1920's. When a motor is operated by sinusoidal power, shaft voltages are caused by alternating flux linkages with the shaft. The linkages are associated with flux unbalance caused by:

- rotor static or dynamic eccentricity
- rotor and stator slotting
- axial cooling holes in the stator and / or rotor laminations
- shaft keyways
- rotor core support arms
- joints between segmental laminations
- directional properties of magnetic materials
- supply unbalance
- transient conditions.ⁱ

Voltages exceeding 300mV require one bearing of the motor to be insulated to prevent circulating current damage to the bearings. (Refer to Fig. 1) Typically this phenomenon only occurs on 500 frame and larger machinesⁱⁱ. Normally the Opposite Drive End (ODE) bearing is chosen. (If the Drive End (DE) is insulated, the driven load can provide an electrical path that completes the loop to allow current to flow.)

PWM drives can cause increased circulating currents to flow due to a high frequency flux produced by common mode currents which link the stator, rotor and bearing loop. This is an inductive rather than capacitive effect.ⁱⁱⁱ Motors become more asymmetrical at high frequencies because the high frequency capacitively coupled currents depend strongly on the location of the

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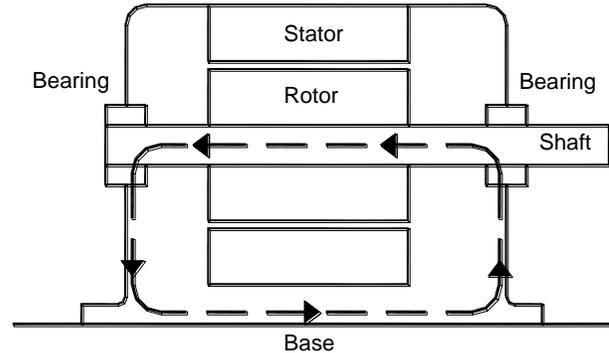


Fig. 1 Showing inductive circulating currents

first few turns within the slot. Since placement of the turns in random wound motors is not well controlled by any manufacturer, even a motor which is symmetrical at low frequencies becomes asymmetrical at high frequencies.^{iv}

In addition to the preceding, PWM drives utilizing Bipolar Junction Transistors (BJT's) or IGBT's can cause Electric Discharge Machining (EDM) currents.^v PWM inverters excite capacitive coupling between the stator windings, the rotor and the stator frame. This common mode current does not circulate but rather travels to ground. Refer to Fig. 2. The path to ground can be through both motor bearings and/or load or auxiliary equipment bearings.

This paper will investigate the phenomenon of induced shaft voltages caused by PWM, AC variable speed drives and some of the methods of mitigating their harmful effects.

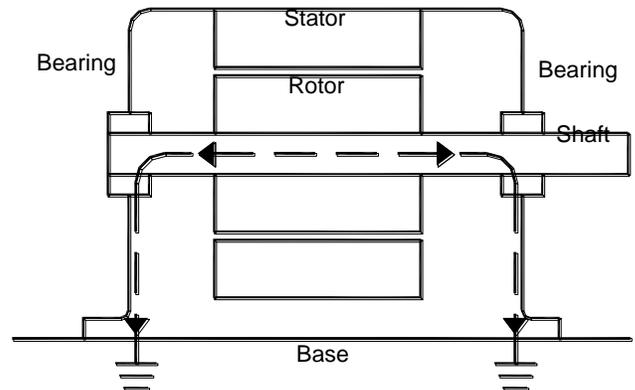


Fig. 2 Showing capacitively coupled current flow

II. RECOGNITION OF SHAFT CURRENT DAMAGE

Ideally, when a bearing fails, the cause of the failure is investigated by examination. Often, bearings are replaced during normal maintenance procedure and the root cause of the problem is not always immediately investigated. This makes elimination of the failure source much more difficult since the equipment is back in service.

Electrical damage to anti-friction bearings primarily appears as fluting. Initially, EDM currents cause permanent microscopic marks in the bearing race. The marking interval is evenly spaced according to the ball spacing. The initial microscopic marks causes slight vibration which is too small to be picked up by vibration analysis equipment. Bearing balls or rollers fall into these microscopic pits in the race's load zone displacing a small portion of lubricant. Slight removal of some of the lubricant reduces the dielectric value allowing a voltage transient to cause a larger current to flow. This damage causes the bearing threshold voltage to be lowered allowing lower level transients to further mark the bearing. Continued deterioration usually occurs at the bottom of the original race markings. This is why fluting marks occur in the same place on the bearing race load zone and why many bearing fluting failures appear the same.^{vi} An example of fluting damage is shown in Fig. 3. (Current damage often initially appears as frosting which looks like a satin finish on the raceways and balls or rollers).

One study^{vii} investigated 1150 ASD powered AC motors in similar clean room applications. The results showed that 25% of motors in operation for less than 18 months had bearing damage caused by electrical discharge. It also showed that motors that had been in operation for greater than 18 months, with an average age of two years had 65% of the bearings damaged due to electrical discharge. Note that clean room applications utilize high carrier frequencies (typically 12KHz or higher) to minimize motor audible noise and run at the same speed continuously.

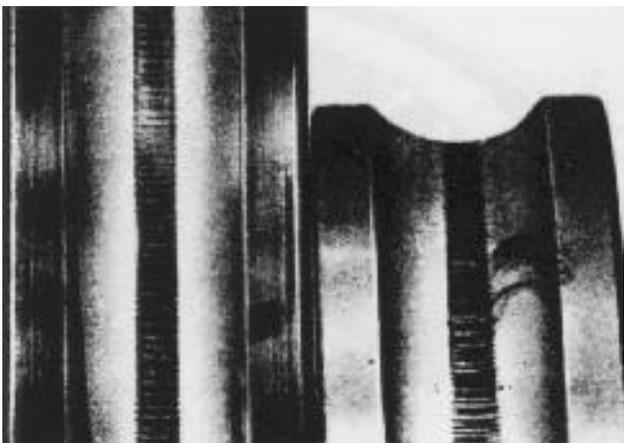


Fig. 3 Typical fluting damage to anti-friction bearings.

Theoretically a periodic square wave is composed of a fundamental frequency plus infinite harmonic frequencies. Because of the very fast rise times of IGBT's, the PWM pulses produced are almost perfect square waves, rich in high frequencies. Since a capacitor is approximately a short circuit at high frequencies, IGBT's induce more capacitively coupled current than slower devices ($I \approx C \times dv/dt$).

Higher carrier frequencies induce more capacitive energy to the rotor and the stator frame since there are more pulses in a given period.

Our field experience has shown that IGBT drives with high carrier frequencies (e.g. above 8 KHz) lead to significantly faster bearing deterioration than lower carrier frequencies.

An example of problems related to ASD's occurred at a large semiconductor manufacturer in the US Pacific North West. After 18 months of operation a pattern of bearing noise and vibration was noticed in a population of approximately 100 motors. The motors were 15 HP, 460V, 900RPM, Open Drip Proof (ODP) powered by high carrier frequency PWM IGBT inverters. The application was plug fan drives in a clean room chip fabrication facility. Even though the noise and vibration were at relatively low levels and the motors continued to operate, it was still considered a failure in this application. Initial inspection verified that bearing damage was a result of shaft current caused by IGBT inverters. The next step was to repair the motors and apply counter-measures to prevent further problems. In an effort to minimize down time and cost, a single insulated bearing bracket was fitted. As discussed earlier, this was the most common solution used on large machines for many years. Insulating one bearing, however, proved to be unsatisfactory because the un-insulated bearing failed at a faster rate. The next measure was to install shaft-grounding brushes. This was also not completely successful due to improper installation. Further, brushes posed maintenance problems and contamination concerns due to the carbon brush material. As a retrofit, insulating one bearing and adding a ground brush was the most practical choice, however, in hindsight, it would have been better to insulate both bearings. At the time, the phenomenon was not as well understood and this preventative measure was not considered.

III. BEARING DAMAGE MECHANISM

The degree of damage caused by EDM currents is dependent on many factors. The contact area, which consists primarily of irregularities on the surfaces of the balls or rollers and races touching each other, determine how much current can flow without causing localized overheating to the bearing assembly. At standstill, there is

significant contact area and low resistance. As the motor speeds up (typically above 10% of rated speed), the bearings “float” on a film of oil. The effective resistance of this film is a function of the film thickness and the type of lubricant.

When a high resistivity grease is used and the bearings are “floating” on the oil film, the equivalent circuit characteristic changes from a resistor to a capacitor. Imperfections on the bearing surfaces occasionally puncture the oil film and discharge the rotor. (Discharges due to metal to metal contact occur at low voltage levels and do very little direct damage to the bearing surface.)

The better the quality of the bearing, the less often these low level discharges occur allowing the rotor to charge for longer periods of time and hence attain higher voltage levels. (Typically, high quality bearings charge as much as 80% of the time due to a uniform oil film. Low quality bearings charge as little as 5% of the time due to frequent metal to metal contact.) If the rotor voltage exceeds the threshold voltage (V_{th}) of the oil films between the balls or rollers and the races of the bearing, the oil film’s dielectric strength is exceeded. At this point destructive EDM currents and arcing occur. It is interesting to note that, the contact time between the balls and outer race is longer than contact time between the balls and inner race, hence the bearing wear from EDM and dv/dt currents is greater in the outer race.^{viii}

The existence of EDM currents with PWM voltage source inverter drives depend on the presence of all of the following conditions:

1. Excitation, which is provided by the source voltage to ground (V_{sg})
2. A capacitive coupling mechanism, between stator and rotor (C_{sr})
3. Sufficient rotor voltage buildup which is dependent on the existence of bearing capacitance (C_b)

There are two basic groups of variables which affect capacitive bearing current:

1. Mechanical Variables – Shaft voltages and bearing currents depend on the existence of C_b . Bearing impedance becomes capacitive only when a lubricating film occurs in the contact regions between the balls or rollers and the raceways. The capacitance is dependent on film thickness which is a function of radial load, velocity, temperature, lubricant dielectric strength and lubricant viscosity. The contact area increases proportional to the bearing load raised to approximately the $1/2$ power.^{ix}
2. System Impedance – The system impedance is represented in Figs.4 and 5 which is composed of the stator winding to frame capacitance (C_{sf}), the stator winding zero sequence impedance (L_o and R_o), rotor to frame capacitance (C_{rf}), C_{sr} and C_b . Z_l accounts for the mechanical and electrical abnormalities and randomness of the bearing.

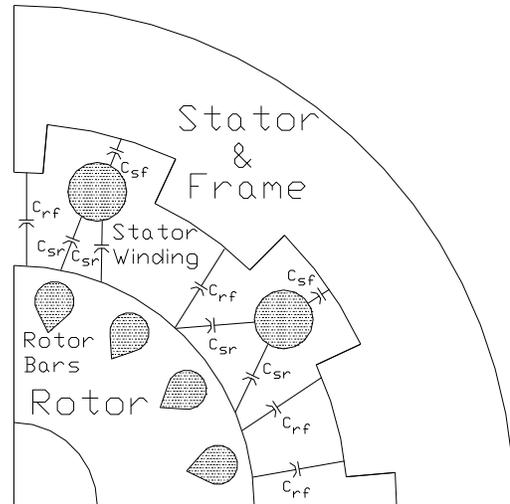


Fig. 4 Pictorial Diagram of Motor Capacitances^v

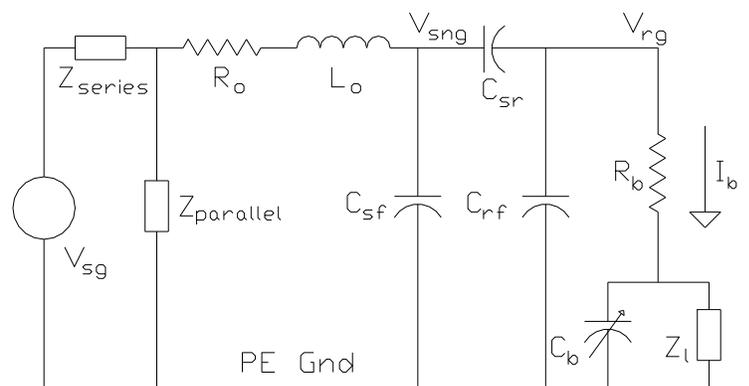


Fig. 5 Common Mode Equivalent Model^v

R_o & L_o are stator winding zero sequence impedance
 V_{sg} is the Source Voltage to Ground
 V_{sng} is Stator Neutral to Ground Voltage
 V_{rg} is Rotor to Ground Voltage
 C_{sf} is Stator Winding to Frame Capacitance
 C_{sr} is Stator Winding to Rotor Capacitance
 C_{rf} is Rotor to Frame Capacitance
 C_b is Bearing Capacitance
 R_b is Bearing Resistance
 Z_l is the variable impedance of the bearing which often is represented as a switch which randomly closes due to quasi-metallic surface contact
 PE Ground is Protective Earth Ground

IV. MITIGATION OF BEARING DAMAGE DUE TO ELECTRICAL DISCHARGE

As mentioned previously, low quality bearings have more contact between balls or rollers and the raceways because of surface irregularities. This increased contact provides less opportunity for damaging EDM currents and arcing to occur. The implication is that a low quality bearing may, given a certain set of circumstances, provide longer life than its higher quality counterpart. This paper is not suggesting that using low quality bearings is a recommended solution.

EDM discharge can be virtually eliminated by providing electrostatic shielding between the rotor and stator. This same concept that has been used for many years in transformers and is referred to as a Faraday shield. One method of accomplishing this is to install a grounded metallic foil tape so that it covers the stator slots and the end turns of the windings. Experiments on an unloaded motor have confirmed that this dramatically reduces the rotor voltage and the dv/dt currents produced through the bearings.^{viii} This solution may not be practical since special motors and spares are required.

Another solution to reduce rotor to ground voltage (V_{rg}) buildup is to utilize a low conductivity grease. The associated drawback is that experience with conductive grease shows that bearing life decreases by a factor of three as compared to utilizing conventional grease.ⁱⁱⁱ One bearing manufacturer suggests that damage due to bearing currents can be considerably reduced by utilizing a low viscosity grease of about 7 centistokes with the addition of graphite to reduce current density. The size of the graphite particles should be of approximately the same thickness as the lubricating film between contact surfaces so that the conducting area is increased, but not so small that the particles are suspended in the oil film.^x (A point of interest is that as bearings age, contaminants from EDM and mechanical wear can significantly change the lubricant's electrical characteristics.)

The most common solution to eliminate bearing currents is to insulate the bearings. Please note that both bearings must be insulated because capacitively induced currents flow to ground. If only one bearing is insulated, all the current will flow through the un-insulated bearing causing a rapid failure. If the insulated bearing solution is chosen, any connected mechanical load must be insulated or the current will flow from the motor shaft through the load bearing(s) to ground or through other connected components such as tachometer bearings. If it is not possible to isolate the mechanical load and connected components, a shaft grounding brush should be added to provide a low impedance path to ground.

A common solution, which has been used with DC machines for many years, is to simply add a shaft grounding brush without insulating the bearings. When adding a grounding brush, it is important that the motor frame be adequately grounded for high frequencies. If not, there is still a potential current path through the load bearing(s).

Induced voltage is created by the steep wave front of each PWM pulse. The faster the rise times, the higher the orders of harmonic currents present. The spectrum ranges from almost DC to as high as 30 MHz for IGBT drives which switch at 0.1 μ s.

High frequency currents only travel on the surface of the grounding conductor. If the ground conductor is long it is very possible that a portion of the current will

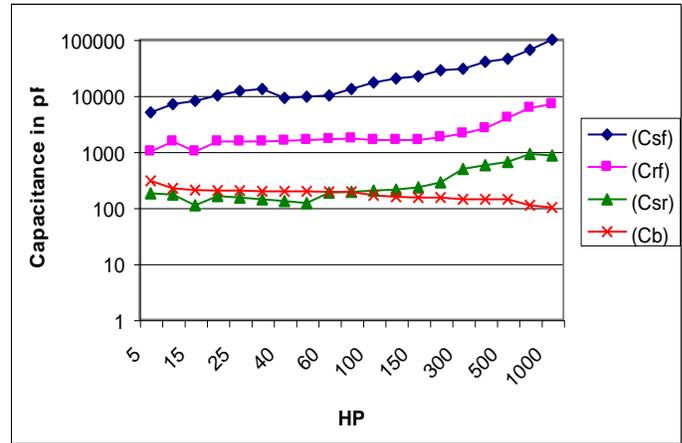


Fig. 6 Relative Motor Capacitance Values^v

flow from the motor frame, through the load bearings to ground.

Effective high frequency grounding can often be accomplished by simply making sure that the motor mounting base is welded to the mechanical load's base. A braided copper grounding strap bonding the motor frame to the mechanical load's base will serve the same purpose. (Sometimes the load is connected to the framework of the building through piping etc. which, to high frequencies, is a lower impedance than a conventional ground wire from the motor frame to the electrical ground.)

As seen in Fig. 6,^{iv} the calculated capacitance between the windings and the stator frame C_{sf} is typically in the order of 30 - 100 times higher than the capacitance between the windings and the rotor C_{sr} . This means that stator coupling currents, though not normally taken into account, are considerably large. If the motor is not adequately grounded for high frequencies or not grounded, these currents can flow through the shaft brush and then through the load bearings. In this case adding a shaft brush can actually increase bearing currents in the driven equipment. Refer to Fig. 7.

Following the same logic, if a shaft brush is not

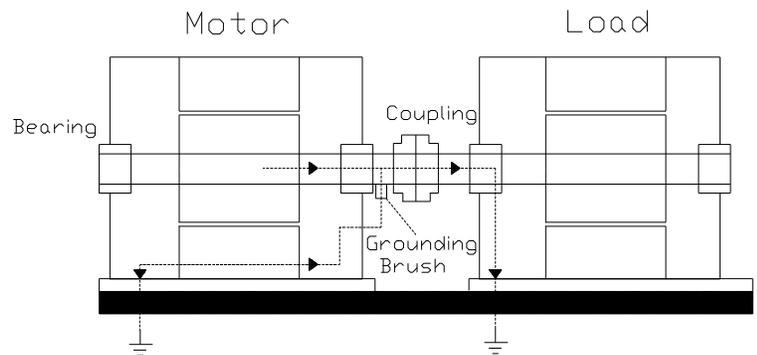


Fig. 7 Showing paths for common mode currents to ground

TABLE I^{xi}
TEST RESULT COMPARISONS FOR A 20HP MOTOR

	Without Filter	Conventional Filter	Proposed Filter
Peak Voltage at Motor Terminal	1260V	750V	730V
Differential mode dv/dt at motor terminals	3037 V/ μ s	1267 V/ μ s	120 V/ μ s
Induced shaft Voltage to Ground (RMS)	864 mV (RMS)	442 mV (RMS)	234 mV (RMS)
Leakage Current to Ground (RMS)	430 mA (RMS)	252 mA (RMS)	132 mA (RMS)
Total Filter Power Loss (Watts)	0 (Watts)	53.5 (Watts)	125 (Watts)

utilized, it is very important that the motor frame is well grounded. This is because the voltage produced by the capacitance between the windings and the stator frame (C_{sf}) may discharge through the motor bearings and to ground through the load bearing(s).

Research is ongoing with soft switching type PWM ASD's as compared to the conventional hard switching types. To date there has been limited success in eliminating bearing discharge. The problem with soft switching and other methods of improving the PWM output waveform is that when viewed phase to phase it looks very clean but phase to ground (common mode) it is not. Common mode noise is what excites the stray capacitance between the stator windings to the rotor and the stator frame causing voltage build up.

The ratio of common mode noise caused by a PWM drive compared to a sine wave is in the order of 10 – 1 or more. The addition of a filter which combines both common mode and differential mode filtering can reduce this ratio by as much as 70%. Refer to Table I for test results completed with a 20HP, 460V, high efficiency induction motor and 480V PWM IGBT inverter utilizing this output filter compared to a conventional filter or no filter. Note that a conventional filter (without the common

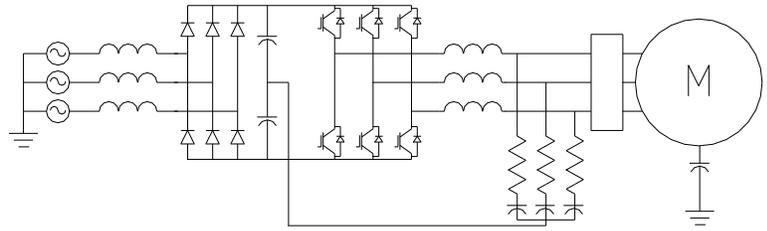


Fig. 8b Proposed output filter to reduce differential mode and common mode dv/dt at the terminals of the motor.^{xi}

mode connection), which has oversized inductance and capacitance such that a near sinusoidal line to line voltage is provided, common mode dv/dt is still high.^{xi}

Figure 8a shows the path of common mode current flow in a typical installation. Figure 8b shows the addition of a common mode filter which connects the wye point of the filter to a “neutral” point on the DC bus. This filter arrangement provides a low impedance path from the output of the ASD back to a neutral point on the DC bus instead of through the motor. (Note that further research has shown that the wye point of the filter can be connected to the Neg. DC bus with similar results.^{xi})

An easily implemented mitigation means is to keep the carrier frequency as low as practical. A value of between 1500 and 3000Hz minimizes the amount of energy transferred to the rotor while maintaining good drive performance.

Output reactors have an effect on the common mode noise generated which can increase the possibility of elevating U_L . This augments the probability of C_b charging thereby causing EDM discharges to occur. RLC type output filters should be considered instead.

Cable length, cable type and grounding arrangement can impact the likelihood of discharge through the motor, load or auxiliary equipment bearings. Refer to Fig. 9a for a simplified physical arrangement

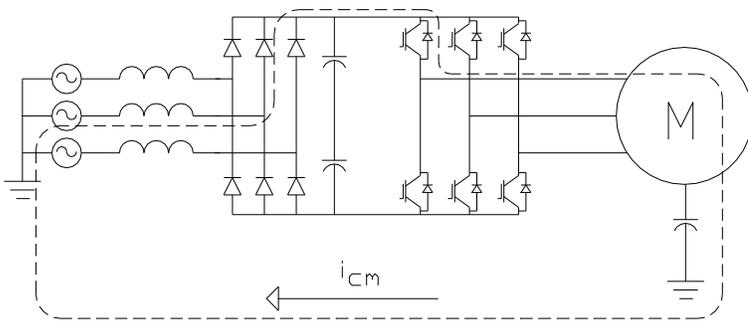


Fig. 8a Showing path of common mode current flow in a conventional ASD and motor.^{xi}

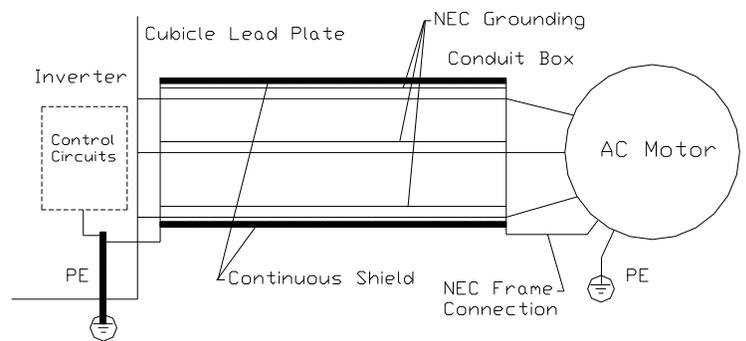


Fig. 9a Cable Configuration Diagram^{xii}

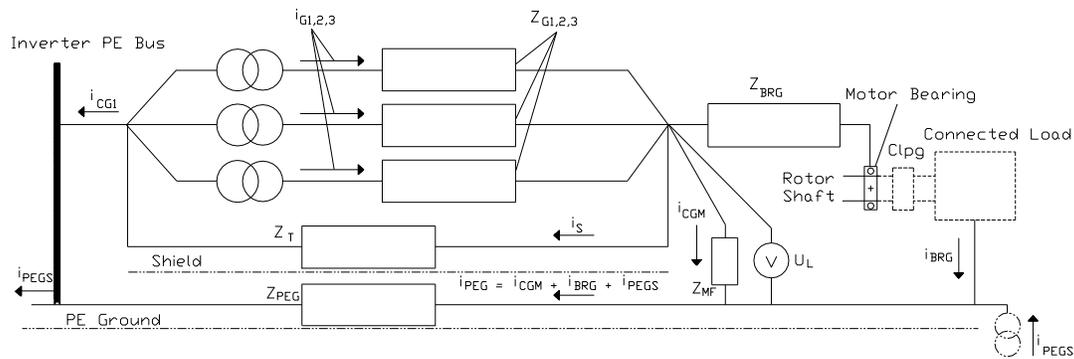


Fig. 9b Cable Equivalent Circuit^{xii}

$Z_{G1,2,3}$ is the common mode impedance in each of phases 1, 2 & 3

Z_{BRG} is the equivalent bearing impedance

Z_S is the shield or sheath impedance

Z_{MF} is the common mode impedance to ground through the motor frame

Z_T is the cable shield or sheath transfer impedance

Z_{PEG} is impedance of the Protective Earth Ground

$i_{G1,2,3}$ is the common mode current in each of phases 1, 2 & 3

i_s is the shield or sheath current

i_{PEG} is the Protective Earth Ground current

i_{CGM} is the common mode current through the motor frame to ground

i_{BRG} is the bearing current to ground

i_{PEGS} is the Protective Earth Ground System current flowing from other parts of the system

U_L is the Transfer Impedance voltage drop

drawing and Fig. 9b for the equivalent schematic of a typical cable. Due to the existence of stray capacitance, cables cause common mode currents to flow. If the motor frame is not grounded (Z_{MF} is open circuited) or if Z_{MF} has a high impedance to high frequencies, the return path for common mode current is the cable shield (or armor).

Shield currents produce a resultant cable transfer impedance (Z_T) voltage drop shown as U_L in Fig. 9b. The higher the value of the cable shield or sheath transfer impedance (Z_T) and of common mode impedance to ground through the motor frame (Z_{MF}), the more likely that discharge will occur through the alternative current return path which is through the PE Ground via the motor and / or load bearings.

Testing has shown that cables which have a continuous shield or continuous armor provided the lowest common mode current plus relatively low frame voltage.^{xii}

The recommended cable for PWM ASD application has six symmetrical conductors, 3 \varnothing and 3 ground conductors) with a continuous corrugated aluminum armor-type sheath. To make sure that the

cable characteristics are fully exploited, proper connectors need to be utilized to maintain low ohmic contact resistance to the armor which essentially becomes a shield.^{xii}

V. CONCLUSIONS

When a bearing fails, especially on a motor being powered by a PWM ASD, the bearing and lubricant should be examined to determine the cause of failure. If the damage is due to EDM, corrective measures should be considered.

As discussed, there are several possible practical solutions to mitigate bearing currents which include:

1. Selecting a carrier frequency which is between 1500 and 3000Hz if practical. This significantly reduces the energy transferred to the rotor.
2. Adding a common mode filter to mitigate common mode noise
3. Insulating both bearings to prevent current flow
4. Adding a shaft grounding brush or brushes to shunt common mode currents (ideally with the ODE bearing being insulated).
5. Making sure that the motor frame is suitably grounded for high frequency currents to prevent stator frame currents flowing through the bearings to the connected load.
6. Changing the cable to the recommended type to minimize the common mode current.
7. As a temporary measure, utilizing conductive grease.

New installations should be designed with the bearing current phenomenon in mind and take into account the issues discussed in this paper. This is particularly important if high carrier frequencies are planned to be used.

ⁱ Pratt, J.W., "Shaft Voltages Caused By Alternating Flux Encircling The Shaft", 'Electrical Machines and Drives', 11-13 September 1995, Conference Publication No. 412 © IEEE, 1995

ⁱⁱ NEMA MG1 Section 31

ⁱⁱⁱ Shaotang Chen, Thomas A Lipo, Donald W. Novotny, "Circulating Type Motor Bearing Current in Inverter Drives", IEEE – IAS Conference Record, 1996, pp 162-167

^{iv} S. Bhattacharya, L. Resta, D.M. Divan, D.W. Novotny, T.A. Lipo, "Experimental Comparison of Motor Bearing Currents with PWM Hard and Soft Switched Voltage Source Inverters", IEEE – IAS Conference Record, 1996, pp 1528-1534

^v Doyle Busse, Jay Erdman, Russel J. Kerkman, Dave Schlegel and Gary Skibinski, "Bearing Currents and Their Relationship to PWM Drives", IECON '95, IEEE 21st Annual Industrial Electronics Conference, Nov 6 – 10, 1995, Vol. 1, pp 698-705 or IEEE Transactions on Power Electronics, Vol. 12, No. 2, March 1997, pp 243-252.

^{vi} J. Alan Lawson, "Motor Bearing Fluting" CH3331-6/93/0000-0032, © 1993 IEEE

^{vii} Shaft Grounding Systems Inc

^{viii} Jay M. Erdman, Russel J. Kerkman, David W. Schlegel and Gary L. Skibinski "Effect of PWM Inverters on AC Motor Bearing Currents and Shaft Voltages" Paper IPCSD 95-63

^{ix} T. Harris, "Rolling Bearing Analysis". New York: Wiley, 1984

^x SKF publication "Passage of Electric Current Through a Rotating Contact"

^{xi} D. Rendusara, P. Enjeti, "New Inverter Output Filter Configuration Reduces Common Mode and Differential Mode dv/dt at the Motor Terminals in PWM Drive Systems", 28th IEEE Power Electronics Specialists Conference, 1997 Vol. 2 pp 1269-1275

^{xii} John M. Bentley, Patrick J. Link, "Evaluation of Motor Power Cables for PWM AC Drives", in Conference Record of 1996 Pulp & Paper Industry Technical Conference, IEEE Industrial Application Society, pp 55-69